

AS1602 — Dual Channel Active EMI & ESD Suppressor for Ethernet Applications

GENERAL DESCRIPTION

The AS1602 is a single-chip, highly integrated CMOS solution for Common Mode (CM) noise suppression & transient voltage protection in Ethernet applications.

The AS1602 EMI/ESD Suppressor has been architected and designed to provide system level ESD protection and EMI suppression in Ethernet products, enabling an easier path to system EMC compliance.

The AS1602 utilizes Akros Silicon's patented Active Choke technology which offers superior Common Mode reduction and immunity compared to passive filtering techniques in Ethernet applications. The adaptive and continuous suppression operates over the entire Ethernet signal bandwidth, and it compensates for many variables that are the source of common mode noise in Ethernet systems. It also improves the differential-to-common mode balance of the system. This enables system designers to meet EMI emissions and EMI immunity requirements from the start of the design.

In addition, the AS1602 includes highly robust ESD/Surge protection diodes to protect the Ethernet PHY from various transient overvoltage events. These are built using Akros' proprietary design and layout techniques to safely deal with very high current densities in a CMOS process without creating voltage and/or thermal overstress that causes damage to the device.

Each AS1602 supports two twisted pair interfaces. 10/100 applications require one AS1602 device while 10/100/1000 applications require two AS1602 devices. The AS1602 is available in a small footprint 16-pin QSOP Reduction of Hazardous Substance (RoHS) compliant package.

TYPICAL APPLICATIONS

- Ethernet systems requiring additional CM suppression to meet EMC Class B emissions or higher EMI immunity requirements and ESD protection
- POE and Non-PoE Ethernet systems
- VoIP Phones, IP Cameras, WAPs, Routers, Switches
- Set Top Boxes, Networked Printers and Appliances, Desktop and Laptop Computers

FEATURES

- Enables system designers to comply with:
 - CISPR22 and FCC Part 15, Class B requirements for Radiated and Conducted Emissions
 - IEC 61000-4-3 requirements for Radiated and Conducted Immunity, Level 3 or higher
 - IEC 61000-4-2 ESD (Air Discharge) of $\pm 25\text{kV}$
 - IEC 61000-4-2 ESD (Contact Discharge) of $\pm 12\text{kV}$
 - Cable Discharge Event (CDE) of $\pm 12\text{kV}$
- Provides up to 10dB of additional common mode noise suppression over the frequency of 1MHz to 125MHz when used with Ethernet magnetics
- Robust built-in ESD suppressors protect the Ethernet PHY and improve system ESD performance
 - JESD22-A114, ESD, HBM of $\pm 8\text{kV}$
 - Interfaces to standard Ethernet transformers and 10/100/1000 Ethernet PHYs
 - Uses a single standard power rail (3.3V or 2.5V)
 - Open drain output stage that can be biased from 1.8V to 3.3V using transformer center-tap supply as needed based on choice of Ethernet PHY.
 - Flow-through routing for ease of board layout
 - Typical power consumption of 90mW.
 - Low power mode available
 - Industrial temperature range (-40°C to 85°C)

SIMPLIFIED APPLICATION DIAGRAM

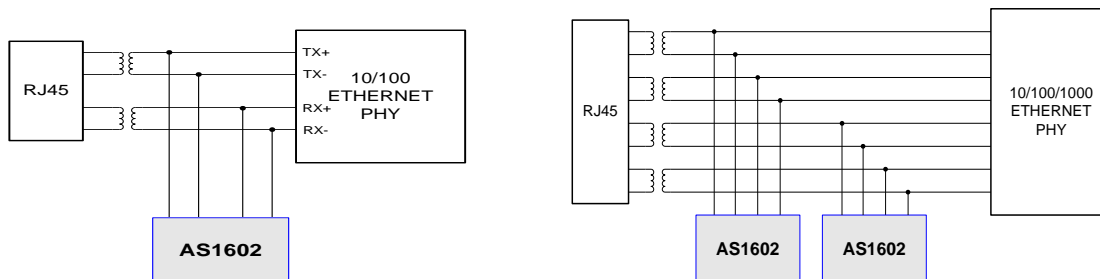


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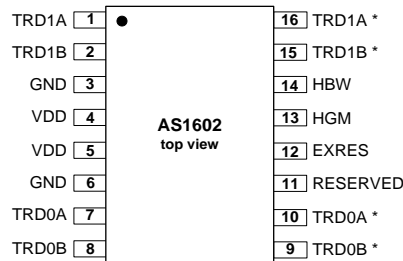
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PIN ASSIGNMENTS AND DESCRIPTIONS

Figure 1 – AS1602 Pin Assignments



* Not connected internally – pin named to enable PCB flow-through routing.

Table 1 – AS1602 Signal Descriptions

Pin	Name	I/o ¹	Description
1, 16	TRD1A	A, OD	PHY media differential pair 1 positive or negative signal. Note that circuit electrical connection is to pin 1 only. Pin 16 is named the same to enable through-routing on the PCB, however it is <i>not</i> connected internally.
2, 15	TRD1B	A, OD	PHY media differential pair 1 positive or negative signal. Note that circuit electrical connection is to pin 2 only. Pin 15 is named the same to enable through-routing on the PCB, however it is <i>not</i> connected internally.
3, 6	GND	P	Board ground - same as Ethernet PHY ground
4, 5	VDD	P	Analog VDD. 3.3V ± 5% or 2.5V ± 5% supply can be used. VDD can be shared with transformer center-tap supply, but is not required. Note that the transformer center-tap supply voltage (VCT) must always be <i>equal to</i> or less than VDD.
7, 10	TRD0A	A, OD	PHY media differential pair 0 positive or negative signal. Note that circuit electrical connection is to pin 7 only. Pin 10 is named the same to enable through-routing on the PCB, however it is <i>not</i> connected internally.
8, 9	TRD0B	A, OD	PHY media differential pair 0 positive or negative signal. Note that circuit electrical connection is to pin 8 only. Pin 9 is named the same to enable through-routing on the PCB, however it is <i>not</i> connected internally.
11	RESERVED	-	Reserved - leave open
12	EXRES	A	External bias resistor. For VDD = 2.5V use 8.06kΩ ±1%. For VDD = 3.3V use 10.7kΩ ±1%. Connect resistor between EXRES pin and GND
13	HGM	DI	Allows trade-off between desired CM rejection performance and device power consumption: FLOATING: Default setting - recommended normal operating condition LOW: Lower power mode setting - reduces CM rejection performance to reduce power consumption. See “Performance/Power Control Pins” for additional details. HIGH: Reserved - should not be used
14	HBW	DI	Allows control of CM rejection performance by changing internal loop bandwidth. May be used in some systems to optimize EMI performance. FLOATING: Default setting - recommended normal operating condition LOW: Reserved - should not be used HIGH: Increases loop bandwidth to increase CM rejection performance. May be used in some systems to optimize EMI performance. See “Performance/Power Control Pins” for additional details.

1. OD = Open Drain, A = Analog signal, DI = 3-state Digital Input signal, P = Power

Table 2 – Absolute Maximum Ratings

Parameter	Max	Unit
PHY media TRDn pins	5.0	Volt
All other pins	3.6	Volt
Supply voltage (VDD)	3.6	Volt
ESD Human Body Model (JESD22-A114) ¹	8	kV
ESD Charged Device Model (JESD22-C101) ¹	500	Volt
System ESD (IEC 61000-4-2: Air Discharge) ²	25	kV
System ESD (IEC 61000-4-2: Contact Discharge) ²	12	kV
Cable Discharge Event (CDE) ²	12	kV
Storage temperature	165	°C
Junction temperature	125	°C

¹ Human Body Model and Charged Device Model ESD limits are specified at the chip level.

² Air Discharge, Contact Discharge, and Cable Discharge Event maximum limits are specified at the system level.

Table 3 – Recommended Operating Conditions

Symbol	Parameter	Min	Typ	Max	Unit
V _{DD}	3.3 V Power supply (VDD)	3.13	3.3	3.47	Volt
	2.5V Power supply (VDD)	2.37	2.5	2.63	Volt
V _{CT}	DC common mode of Ethernet signals (transformer center-tap supply - see note 1)	1.8 -5%		3.3 +5%	Volt
T _A	Operating temperature range	-40	–	85	°C

¹ The transformer center-tap supply voltage (V_{CT}) must always be *equal to or less than* V_{DD}.

Table 4 – Electrical Characteristics

Parameter	Test conditions	Min	Typ	Max	Unit
Differential signal insertion loss	0.1 MHz - 100 MHz	–	0	0.2	dB
Differential output capacitance (between TRDnA and TRDnB pins)	–	–	4.5	–	pF
Common mode PHY noise rejection (25Ω CM impedance)	at 1 MHz	30	–	–	dB
	at 100 MHz	10	–	–	dB
Common mode PHY noise rejection additive to transformer ³	At 100 MHz	10	–	–	dB
Differential IM3	60 MHz full scale with 100Ω differential resistor across terminals	–	-40	–	dB
Common mode-to-differential mode imbalance	1 MHz - 100 MHz	–	–	–	dB
Power dissipation	Default mode: HGM = floating V _{DD} = 2.5V, V _{CT} = 1.8V	–	90	130 ²	mW
	Low Power Mode: HGM = Low V _{DD} = 2.5V, V _{CT} = 1.8V	–	55	80 ²	

¹ Typical specifications are for T_a = +25°C and V_{DD} = 2.5V. Typical specifications not 100% tested. Performance guaranteed by design and/or other correlation methods.

² Maximum power consumption is with +5% supplies and over all process/temperature range.

³ This measurement is done with typical Ethernet transformer. Unless otherwise specified, all other performance measurements are for standalone AS1602 without Ethernet transformer.

Table 5 – Digital input Characteristics

Symbol	Parameter	Test conditions	Min	Max	Unit
V _{IH}	Input HIGH level	V _{CC} = 2.5V	1.78	3.47 (3.3V + 5%)	V
		V _{CC} = 3.3V	2.0	3.47 (3.3V + 5%)	
V _{IL}	Input LOW level	V _{CC} = 2.5V	-0.30	0.5	V
		V _{CC} = 3.3V	-0.30	0.5	
I _{IL} , I _{IH}	Input leakage, LOW and HIGH levels	V _{CC} = 2.5V	±25	±40	μA
		V _{CC} = 3.3V	±33	±53	

TYPICAL PERFORMANCE CHARACTERISTICS

Figure 2 – Common Mode Rejection with Typical 2-Core PoE Compatible Transformer using 3-Wire PHY-Side Choke

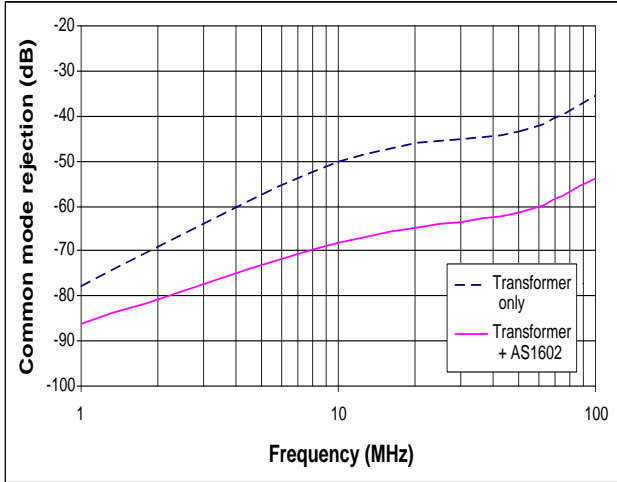


Figure 3 – Common Mode Rejection with Typical 3-Core PoE Compatible Transformer with Media Side Chokes

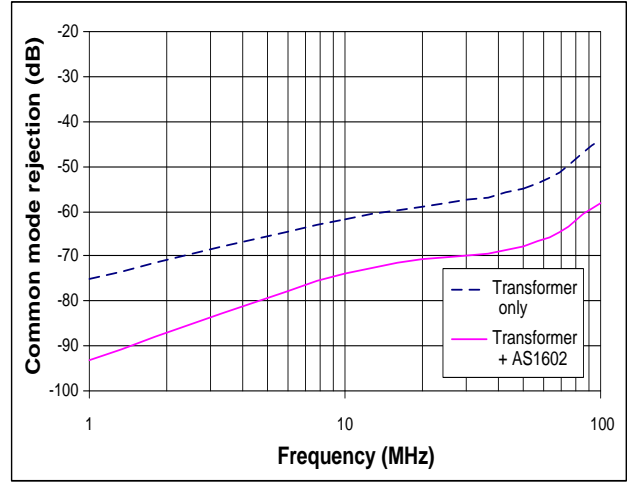


Figure 4 – DM-CM Conversion in Receiving Data from UTP with Typical 2-Core PoE Compatible Transformers using 3-Wire PHY-Side Choke

Figure 5 – AS1602 Return Loss with Typical 3-Core PoE Compatible Transformer with Media Side Chokes

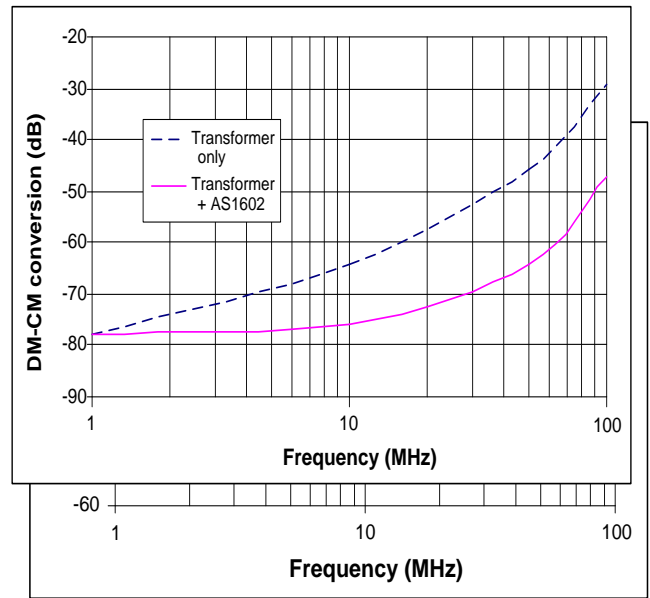
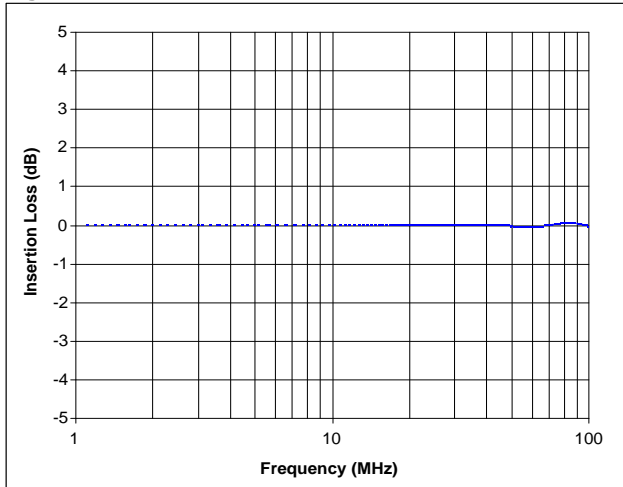


Figure 6 – AS1602 Insertion Loss



FUNCTIONAL DESCRIPTION

Common Mode Noise

The AS1602 compensates for many variables that are the source of common mode noise in Ethernet systems. These include: common mode noise from Ethernet PHY DACs and drivers, variations in Ethernet line signals caused by transformer and passive component mismatches, variations in PC board designs, and different vendor PHY designs.

Most Ethernet PHYs use switched mode D/A converters and drivers to transmit complex modulated signals on the twisted pair. There are inherent common mode switching transients that occur in output driver designs. Switched mode drivers coupled with leakage inductance of the line transformer creates a common mode kick. This common mode noise is coupled through to the twisted pair line via transformer interwinding capacitance. This can result in a significant EMI emissions issue for Ethernet systems that desire good margin from CISPR22 and FCC part 15 Class B specifications.

Ethernet transformers provide common mode rejection through PHY-side autoformer and in-line chokes. Mismatches in the autoformer inductance can lead to differential-to-common mode conversion. Parasitic of the autoformer also limit common mode rejection capability. This residual common mode noise is coupled to the line side through interwinding capacitance.

Additional differential to common mode conversion issues come from board layout mismatches and other passive component mismatches on the PCB. Common mode imbalance at the input of the system can create common mode to differential conversion and significantly impact EMI immunity performance.

As a result, meeting system level EMI requirements is very challenging.

Active Choke Technology

The AS1602 incorporates Akros Silicon’s patented Active Choke technology. The AS1602 is placed between the Ethernet PHY and the line transformer. Its adaptive circuitry monitors both the positive and negative signals of each differential pair and shunts any common mode noise to ground. As a result, noise is prevented from getting through the transformer to the unshielded twisted-pair (UTP).

The AS1602 preserves the integrity of the Ethernet signal by providing very low common mode impedance while maintaining high differential impedance and low differential capacitive loading.

The AS1602 is designed to suppress any common mode noise present in the Ethernet signal pair, irrespective of whether it originates from the PHY side or the UTP side. It provides immunity against large signal common mode events by shunting the common mode signal currents back to ground. Therefore, the Ethernet PHY receiver does not see any common mode movement, enabling it to maintain good link performance.

As shown in Figure 7, the output stage of the AS1602 uses open-drain drivers similar to the most common Ethernet PHYs. The AS1602 can tolerate output DC common mode on its TRD pins within a range of 1.8V to 3.3V. This allows the AS1602 to work seamlessly with any Ethernet PHY and transformer that uses the center-tap to DC-bias the PHY.

The AS1602 architecture uses high precision active analog circuitry which suppresses only the common mode components of the disturbances without impacting the differential signal. This technology provides superior differential-to-common mode impedance balance over traditional transformer solutions.

The AS1602 provides low differential capacitive loading to the Ethernet signals and meets all the common mode-to-differential balance and distortion requirements per the 10/100/1000 Base-T section of the IEEE® 802.3 Ethernet specification.

Transient Threats

Silicon-based electronic devices need protection from spurious events that exert voltage/current stresses exceeding the normal operating limits of the devices. Such events are generally classified as ESD (Electro-Static Discharge) events.

ESD strikes have two modes: differential mode strikes and common mode strikes. Differential mode strikes are strikes which are applied to one pin at a time. These strikes go through the transformer and are the hardest to protect against. Contact/Air Discharge and Cable Discharge event (CDE) can be differential strikes. CDE has become more important in recent years due to advent of cat-5 and better cables. Ethernet cables are frequently dragged across plastic/carpet flooring or through ducts. Charge can accumulate on the cable through tribo-electric effect or induction. The newer cat-5 (or better) cables can hold the charge for a longer period of time relative to older cat1-4 cables. As PHY transceivers have been scaling to finer geometry with lower breakdown voltages, their ability to tolerate large cable discharge events is also limited. Plugging a charged cable into RJ45 jack can lead to an over-voltage stress that can damage sensitive PHY devices.

Common mode strikes normally include EFTB (Electrical Fast Transient/Burst) and surge. The first level of protection of common mode strikes is provided by the isolation of the Ethernet transformer.

Both differential mode and common mode strikes have been the major issues for Ethernet equipment. These transient threats can very easily damage the Ethernet PHYs and can also damage the Ethernet transformer if appropriate design precautions are not taken.

AKROS ESD PROTECTION DESIGN

The AS1602 provides transient protection for both differential mode and common mode ESD events through proprietary CMOS design and layout techniques for low capacitive loading and low inductive effects in protection diodes.

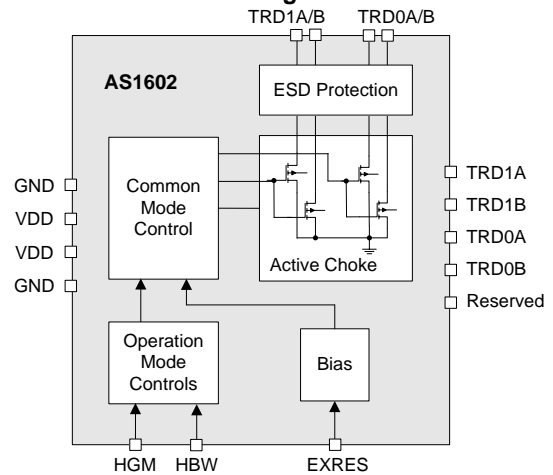
Positive ESD strikes are handled through a stack of diodes that are biased to ensure operation with wide range of V_{DD} and V_{CT} supply voltages and prevent clipping/degradation of Ethernet signals.

Negative ESD strikes are handled through output NMOS bulk diodes, which are also hardened to handle large surge energy. Design prevents snapback of the PHY output NMOS devices by fast turn-on and efficient energy dissipation through ESD structures.

Proprietary design techniques allow the design to handle both moderate energy/fast transients that create high thermal stress as well as high energy/slow transient events that create tremendous voltage overstress.

At the system level, the AS1602 can tolerate over $\pm 25kV$ of IEC61000-4-2 air discharge strikes and $\pm 12kV$ of Cable Discharge Events, and provides protection for Ethernet PHY.

Figure 7 – AS1602 Block Diagram



Performance/Power Control Pins

EMI performance varies significantly from system to system. The AS1602 provides two digital pins (HGM and HBW) that allow various levels of hardware control and performance optimization of the device. The HGM and HBW are 3-input state pins (Low, High and Floating). The Floating state is detected internally by the device as a valid state. For default operation, both pins should be left floating. These pins can be driven by another digital IC with tri-state able outputs or tied in hardware to configure the desired operating mode.

HGM Pin

The default configuration is to leave the HGM pin floating which provides a trade-off between CM rejection performance and device power consumption.

If less rejection is needed in the Ethernet path, the HGM pin should be pulled LOW to select the Low Power Mode. In Low Power Mode, device power consumption is reduced by about 40% and CMRR performance is reduced by approximately 1dB at 100MHz.

HBW Pin

Setting the HBW pin HIGH selects HBW mode which allows system designers another dimension to improve performance by increasing the bandwidth of the internal active circuits. In some systems, this can provide additional performance benefits.

There is a broad range of line transformers that can be used in Ethernet applications. Setting HBW mode may cause oscillations in some systems depending on transformer parasitic. Therefore, HBW mode should only be used if the performance improvement of the default setting (HBW pin = Floating) is not adequate. Ethernet electrical performance should be tested sufficiently to ensure that no oscillatory behavior is seen when using the HBW mode setting.

HBW mode will typically improve the CMRR performance by approximately 1dB at 100MHz.

Note that the HBW mode can be used in conjunction with the HGM pin's Low Power mode. Additional performance/power modulation can be achieved by changing the external bias resistor value (EXRES). For detailed characterization across operating modes, please refer to the relevant application note document.

SYSTEM DESIGN & LAYOUT CONSIDERATIONS

AS1602 Placement

The AS1602 is designed to mate closely with

Ethernet magnetics for ease of board design. The AS1602 should be placed physically near the Ethernet transformer or integrated magnetics jack. Placement on same side of the PC-board as the transformer and the PHY is recommended to avoid vias on the Ethernet traces. The AS1602 should be placed such that pin 1 is facing the transformer or integrated magnetics jack.

Ethernet Signal Polarity Interchangeability

The AS1602 extracts only the common mode component of the Ethernet signals for processing. As a result, the polarity of the Ethernet signals to the AS1602 does not matter. This allows system designers to connect either the positive or negative polarity Ethernet signals to either the TRDnA or TRDnB pins - whichever is convenient for board routing and avoids the necessity of swapping the signals and the use of via insertions.

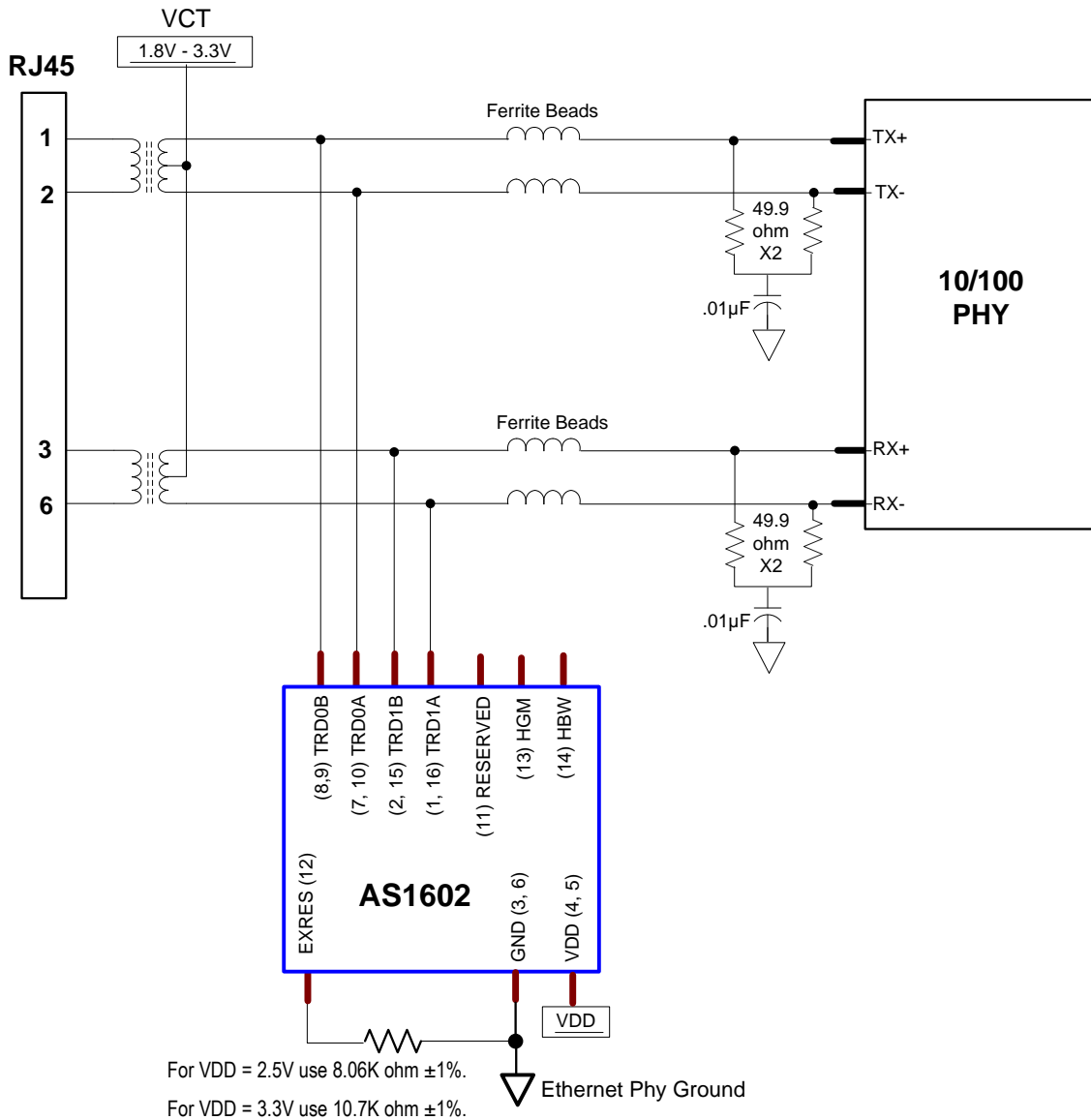
ETHERNET SIGNAL FLOW-THROUGH ROUTING

AS1602 is packaged to allow flow-through routing of the Ethernet differential pairs. Only one side of the package's Ethernet pins (1, 2, 7 and 8) are internally electrically connected to the CM rejection circuitry. Mirrored pins (16, 15, 10 and 9 respectively) are given the same name to enable easy flow-through routing with PC-board design tools.

For a more detailed discussion on designing with the AS1602 and layout considerations, please reference available design guide and application notes.

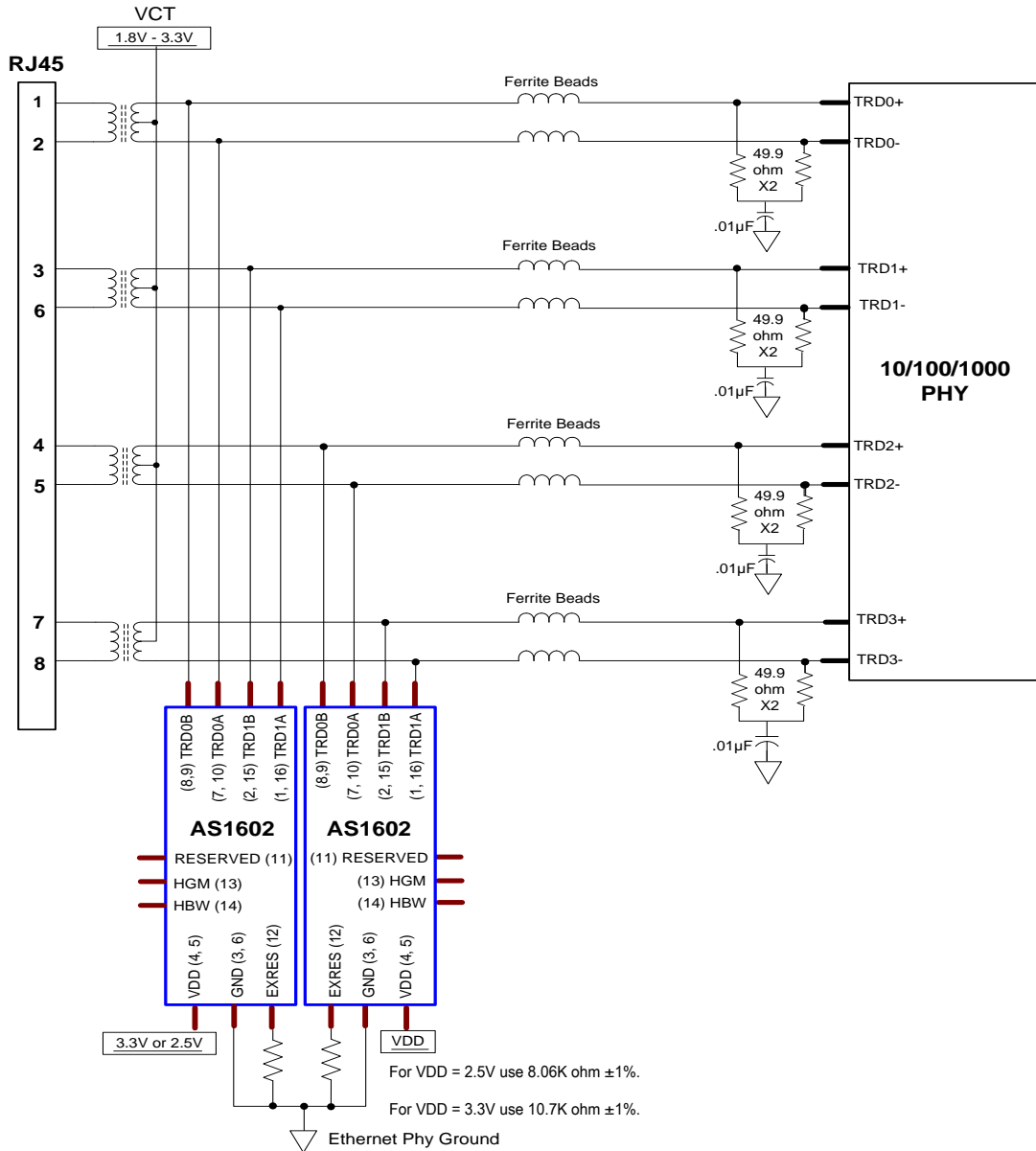
APPLICATION INFORMATION

Figure 8 – Typical FE Application Circuit



NOTE: This diagram is preliminary and subject to change at any time. It may not contain all information necessary to implement the application it represents. Please refer to application notes for details.

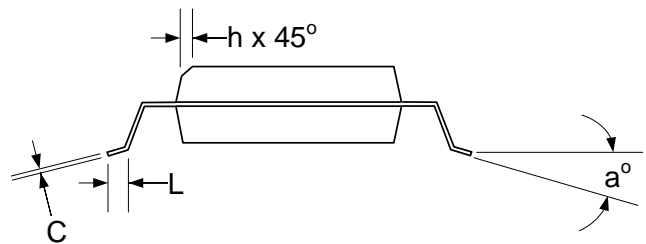
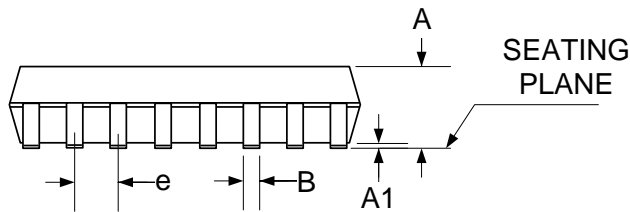
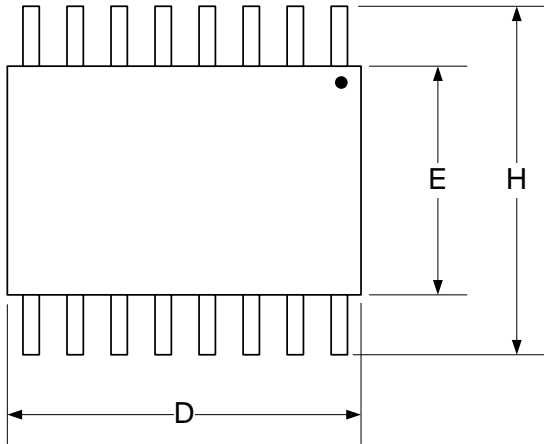
Figure 9 – Typical GE Application Circuit



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PACKAGE SPECIFICATIONS

Figure 10 – 16-Pin QSOP (150 Mil) Dimensions



Symbol	Inches		Millimeters	
	MIN	MAX	MIN	MAX
A	0.060	0.068	1.35	1.75
A1	0.004	0.008	0.101	0.203
B	0.009	0.012	0.203	0.305
C	0.007	0.010	0.178	0.254
D	0.188	0.197	4.80	5.00
E	0.150	0.157	3.81	3.99
e	0.025 BSC		0.635 BSC	
H	0.230	0.244	5.84	6.20
h	0.010	0.016	0.254	0.38
L	0.023	0.029	0.58	0.74
a°	0°	8°	0°	8°

NOTES:

1. Lead width and lead thickness exclusive of solder plate.
2. Package outline exclusive of mold flashes and burr dimensions.
3. Allowable mold flash is 5 Mils per side.
4. Lead coplanarity is 0.003 inch maximum.



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